Design and Identification of Wireless Transmitters for a Low-power and Secure Internet of Things

WENQING YAN
Abstract


Wireless communication is a key enabler for connecting billions of Internet of Things devices. For networked embedded devices operating on limited energy resources, wireless communication dominates the power consumption. Moreover, as networked devices increasingly handle sensitive data, security concerns in wireless communication are continuously expanding. This dissertation develops novel solutions for low-power and secure wireless communication.

Wireless transmitters consist of a series of steps, involving both analog and digital components, each playing a distinct role in the transmit chain. Conventional transmitters employ power-hungry analog components, leading to power consumption on the order of milliwatt. Backscatter transmitters significantly reduce communication power consumption to levels well below one milliwatt. This remarkable power efficiency is achieved by offloading power-hungry components to an external carrier emitter. However, backscatter transmitters encounter challenges in applications that demand medium to long communication range, because they rely heavily on powerful emitters in their proximity for an effective communication range. Instead of removing power-hungry components, our solution integrates the functions of these components into a low-power design. While still requiring an emitter, our transmitter does not reflect the carrier signal. Instead, we utilize a weak carrier signal to stabilize the transmitter, allowing a communication range of over one hundred meters even when the emitter is far away. This contribution takes a step forward in moving low-power communication beyond backscatter.

Passive radiometric fingerprinting leverages imperfections of hardware components to identify and authenticate transmitters. Its passive nature fits well to secure low-power transmitters operating within constrained resources. To enhance the viability of radiometric fingerprinting, we make three contributions in this dissertation to facilitate its widespread deployment. First, compared to conventional radios, low-power backscatter communication has a fundamentally different composition of hardware components in its transmit chain. In our work, we decompose fingerprints in a backscatter system for dual identification of tags and emitters. Beyond security purposes, recognizing the emitter embeds a notion of locality, enabling fingerprinting usage in backscatter network management tasks such as emitter coordination. Second, the dynamic nature of real-world wireless channels significantly impacts the robustness of fingerprinting. We decompose channel impacts and develop a hybrid system. This system employs pertinent strategies for different channel factors, ensuring reliable performance across complex wireless conditions. Lastly, based on the understanding of components' contributions to the transmit chain, we design a lightweight fingerprinting system. We demonstrate a complete implementation seamlessly integrated within the constraints of a single low-cost off-the-shelf chip. This contribution simplifies the conventionally bulky setup using sophisticated signal acquisition equipment and dedicated computer processing resources, which facilitates the practical deployment of fingerprinting on low-cost embedded devices.

Keywords: Wireless transmitters, Wireless embedded systems, Physical-layer security, Radiometric fingerprinting, Radio frequency fingerprinting, Identification, Authentication, Backscatter communication, Internet of Things

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Dedicated to my parents and dear Ning
List of Papers

This thesis is based on the following papers, which are referred to in the text by their Roman numerals.

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* Co-primary authors contributed equally to the work.
List of Papers not Included in the Dissertation


List of Posters and Demos


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December 2023

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List of Acronyms

<table>
<thead>
<tr>
<th>Acronym</th>
<th>Full Form</th>
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<tbody>
<tr>
<td>AWGN</td>
<td>Additive White Gaussian Noise</td>
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<tr>
<td>COTS</td>
<td>Commercially Off The Shelf</td>
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<tr>
<td>DC</td>
<td>Direct Current</td>
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<tr>
<td>DFE</td>
<td>Direction Finding Extension</td>
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<td>DSP</td>
<td>Digital Signal Processing</td>
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<tr>
<td>FPGA</td>
<td>Field Programmable Gate Array</td>
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<td>IC</td>
<td>Integrated Circuit</td>
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<tr>
<td>IDS</td>
<td>Intrusion Detection System</td>
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<tr>
<td>IF</td>
<td>Intermediate Frequency</td>
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<tr>
<td>IoT</td>
<td>Internet of Things</td>
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<tr>
<td>kNN</td>
<td>K Nearest Neighbor</td>
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<tr>
<td>LiFi</td>
<td>Light Fidelity</td>
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<tr>
<td>LO</td>
<td>Local Oscillator</td>
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<tr>
<td>LOS</td>
<td>Line of Sight</td>
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<tr>
<td>MCU</td>
<td>Microcontroller Unit</td>
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<td>ML</td>
<td>Machine Learning</td>
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<tr>
<td>NLOS</td>
<td>Non Line of Sight</td>
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<tr>
<td>NN</td>
<td>Neural Network</td>
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<tr>
<td>OvO</td>
<td>One vs One</td>
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<tr>
<td>OvR</td>
<td>One vs Rest</td>
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<tr>
<td>RF</td>
<td>Radio Frequency</td>
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<td>RFC</td>
<td>Random Forest</td>
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<td>RFID</td>
<td>Radio Frequency Identification</td>
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<tr>
<td>RNR</td>
<td>Region of Negative Resistance</td>
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<tr>
<td>RSSI</td>
<td>Received Signal Strength Indicator</td>
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<tr>
<td>SDR</td>
<td>Software Defined Radio</td>
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<tr>
<td>SNR</td>
<td>Signal to Noise Ratio</td>
</tr>
<tr>
<td>SoC</td>
<td>System on Chip</td>
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<tr>
<td>SoM</td>
<td>Self-oscillating Mixer</td>
</tr>
<tr>
<td>SVM</td>
<td>Support Vector Machine</td>
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<tr>
<td>TDO</td>
<td>Tunnel Diode Oscillator</td>
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Part I:
Dissertation Summary
1. Introduction

The rapid evolution of the Internet of Things (IoT) has enriched our daily lives, offering expansive applications in healthcare, remote collaboration, and interactive entertainment. The foundational building block of the IoT vision is networked embedded systems, forming an interconnected and intelligent ecosystem. Considering deployment requirements, these embedded devices are typically constrained and operate within limited energy resources, relying on small batteries or energy harvested from the environment. Examples expand from lightweight sensing platforms dispersible by the wind [1] to micro-sensors implanted in the human body [2].

The cornerstone of networked embedded systems is wireless communication. The rising ubiquity of small embedded devices poses a pressing need for developing low-power wireless communication solutions designed for limited power budgets. Moreover, as devices increasingly handle sensitive data, such as health information and biometric data, the importance of secure wireless communication systems is steadily growing. A fundamental challenge arises How to provide low-power and secure wireless communication?

The primary objective of wireless communication is to convey information between points that are not physically connected. As illustrated in Figure 1.1, a wireless communication system consists of a data source, a transmitter encoding the data into a wireless signal, a communication channel through which the signal propagates, a receiver decoding the signal, and a data sink. In the context of this dissertation, radio frequency (RF) wireless communication using electromagnetic waves is considered.

A wireless transmitter consists of three essential blocks to generate electromagnetic waves, as illustrated in Figure 1.2. The protocol stack is essential for formatting data and handling communication tasks such as encoding, error correction, and encryption. It ensures efficient, reliable, and secure transmission of data across networks. The modulator is where the data is modulated onto a low-frequency electrical signal by altering its amplitude, phase, or frequency. Following the modulation, the information-bearing signal passes the RF stage, where it is first mixed with a high-frequency carrier and upconverted to the desired RF frequency band. Then, the electrical signal is amplified and converted to an electromagnetic wave via an antenna. In conventional digital communication systems, the protocol stack and modulator are implemented using digital processing components, while the RF stage predominantly employs analog components such as oscillators, RF mixers, filters, and amplifiers.
Backscatter offers a promising approach to designing low-power transmitters, significantly reducing the power consumption of the communication from milliwatts to microwatts [5]. Backscatter systems decompose tasks in the RF stage and deploy them on two devices: The emitter handles the power-hungry task of high-frequency carrier generation. The passive tag, serving as the data source, exclusively performs low-power modulation operations by reflecting the external carrier signal.

Despite rapid advancements in backscatter technology, there has been no large-scale adoption due to several reasons. One primary concern is the limited communication range in comparison to active radios. In industrial monitoring and agricultural management scenarios, a communication range of tens to hundreds of meters is essential for effective wireless connectivity. To achieve a long communication range between the tag and the receiver, existing backscatter tags heavily depend on the emitter, requiring proximity (a few meters) to a strong emitter. This requirement significantly restricts their deployments in practical scenarios. Additionally, backscatter tags modulate external carrier signals by toggling the antenna terminal using an RF switch in accordance with a digital signal. Due to the usage of digital signals, the spectral efficiency of the reflected signal is inherently poor [6], which complicates the network formation and prevents tags from coexisting with other devices.

The first part of my dissertation introduces a novel low-power transmitter architecture to alleviate challenges inherent in backscatter. Instead of the decomposition strategy used in backscatter, the transmitter integrates the power-hungry tasks of high-frequency carrier generation and the mixing in a single power-efficient step. Our design trades stability for low power consumption. To improve the communication quality, we deploy a weak emitter to stabilize the transmitter. We rethink the asymmetric roles of the emitter and transmitter and weaken the dependence between them. Operating within the same low-power budget as a backscatter tag, our transmitter attains a long communicat-
tion range (more than one hundred meters), even when the emitter is located far away (exceeds one hundred meters).

The increasing trend towards low-power and miniaturized networked embedded devices motivates a reconsideration of how to secure wireless communication systems. Identification and authentication are essential, ensuring that only authenticated devices can connect and exchange data over the wireless network. Traditional methods often apply resource-intensive cryptographic algorithms. Although advancements in lightweight cryptographic solutions and hardware acceleration are promising, adopting such methods puts additional demands on the limited resources available on embedded devices.

Radiometric fingerprinting identifies wireless devices utilizing unique characteristics inherent in the radiated electromagnetic waves. Electronic components in the transmit chain exhibit inconsistent performance due to manufacturing variances [7]. As a result, the signal emitted from the transmitter does not just carry the primary data but also the intrinsic fingerprint of the transmitter hardware. By examining such hardware-specific fingerprints, the receiver can identify the transmitting device. The passive nature of such mechanisms entirely relieves the overhead of identification and authentication functionalities from the transmitter side. The second part of the dissertation focuses on radiometric fingerprinting, covering both active commodity radios and backscatter radios.

The propagation of electromagnetic waves from a transmitter to a receiver is influenced by their distance and environmental factors. Consequently, the received signal is affected not only by the hardware-specific fingerprint but also by the characteristics of the wireless channel. These channel characteristics can influence device identification in both positive and negative ways. From one perspective, channel characteristics can serve as spatial fingerprints to locate and potentially identify the transmitter. Wireless channel properties often vary in real-world situations due to user movements and environment dynamics. From another perspective, the transient effects of the wireless channel can degrade the quality of hardware-specific fingerprints, making it difficult to distinguish individual transmitters. To address this, this dissertation introduces methods to enhance the robustness of radiometric fingerprinting, ensuring reliable transmitter identification.

1.1 Research Questions

My work centers around advancing low-power and secure wireless communication, branching into two directions. The first direction explores how to design low-power transmitters, extending beyond backscatter techniques. The second direction investigates how to apply radiometric fingerprinting techniques for robust and efficient transmitter identification in practical deployments.
1.1.1 Low-power Wireless Transmitters

The vision of IoT is built on embedded systems interconnected through wireless communication. Backscatter technology emerges as a promising solution to enable low-power ubiquitous connectivity, consuming power at least two orders of magnitude lower than commodity active transmitters (see the comparison in Chapter 3).

However, backscatter has not yet achieved widespread adoption among IoT application scenarios. A critical challenge lies in its limited communication range [5], which can be attributed to two main factors. First, the signal undergoes dual path loss, initially during its propagation from the carrier emitter to the tag, and subsequently from the tag to the receiver. As the distance between the carrier emitter and the tag increases, the tag communication range drops significantly [8, 9, 10]. Second, during the backscatter reflection operation, the signal experiences return loss at the tag. Consequently, to compensate for the losses, emitters radiate strong signals [11, 12, 13]. Despite this, a strong bond exists between the tag and the external carrier emitter. For a long communication range of tens to hundreds of meters, the tags need to be located in close proximity to an emitter [9, 10, 14, 15].

Furthermore, backscatter tags modulate information by toggling RF switches among antenna terminals with a digital baseband signal, producing unwanted harmonics and out-of-band interference. While solutions like adjusting antenna impedance to approximate analog baseband signal operations exist, they add complexity to the backscatter design [6, 12, 14, 16, 17]. These challenges lead to the following question: Instead of the reflection-based backscatter mechanism, can we design a transmitter with a power budget similar to a backscatter transmitter while offering enhanced communication range and spectral efficiency?

1.1.2 Wireless Transmitter Identification with Radiometric Fingerprints

Hardware-specific fingerprints embedded in wireless signals are used to identify transmitters. Passive radiometric fingerprinting methods are suitable for identifying resource-constrained transmitters, adding a valuable layer of security to wireless communication systems. Such methods can also be integrated with cryptographic techniques to enhance existing security mechanisms. Although radiometric fingerprinting has proven effective, it has not yet achieved widespread deployment.

In real-world scenarios, complex wireless channels challenge the robustness of fingerprinting [18, 19, 20, 21]. In practice, the received signals are compounds of the hardware-specific fingerprints and the transient impact of the wireless channel. It is challenging to perfectly separate them, particularly when the impact of the wireless channel dominates and hides the transmitter’s
fingerprints [20]. Therefore, we ask: How can we improve the robustness of a radiometric fingerprinting system under diverse wireless channels?

There is also an emergence of new low-power transmitter designs, i.e., backscatter technology, with fundamentally different compositions of the electronic components in the transmit chain. We need to understand How do the components contribute to fingerprints in backscatter systems? Is it possible to identify both the carrier emitter and the backscatter tag?

Finally, an important aspect of widespread deployment is enabling fingerprinting on embedded devices with limited resources. Most existing system implementations demand expensive equipment to conduct high-resolution signal acquisition [22, 23, 24], dedicated processing operations [25] and substantial computational tasks [19, 21, 26]. With this in mind, we ask the question: Is it feasible to implement a complete radiometric fingerprinting system on a low-cost and low-power system on chip (SoC)?

1.2 Contributions

In general, my dissertation delves into low-power and secure wireless communication, and contributes to the state of the art in three notable ways. The questions formulated in Section 1.1 are addressed in six papers.

**Design Low-power Wireless Transmitter beyond Backscatter**

We propose a low-power tunnel diode-based self-oscillating mixer (SoM) transmitter architecture (Paper I). The transmitter tag locally generates a weak and unstable high-frequency carrier signal and mixes it with an analog baseband signal in a single power-efficient SoM step. The mixing is accomplished efficiently using the nonlinearity of the tunnel diode, which enables the use of analog baseband signals and prevents the generation of unwanted harmonics. Our transmitter trades stability for low power consumption, which is sidestepped by using an external carrier via the injection-locking phenomenon. Although the system requires an emitter, it fundamentally differs from backscatter by not reflecting the carrier signal. Given that a weak carrier can stabilize the transmitter, the emitter can be located significantly away from the transmitter without compromising the communication range. Expanding on this design, we demonstrate an application of a low-power bridge to replicate visible light communication signals onto radio waves, which allows widespread IoT devices to receive light signals with their existing radio transceivers (Paper II).

**Relation between Transmitter Components and Fingerprint Features**

We attribute fingerprints back to imperfect hardware components and establish a systematic understanding of how imperfections influence fingerprinting features. In the context of backscatter systems, we define fingerprinting features based on the role and characteristics of critical hardware components.
within the backscatter architecture. We emphasize the separation of the carrier and emitter and achieve accurate identification of both tags and carrier emitters, even in complex scenarios with multiple tags sharing several carrier emitters (Paper V). Based on the understanding of the transmit chain, we engineer fingerprint features and integrate their extraction in a coherent receiver architecture, enabling an efficient implementation on embedded devices. We demonstrate a complete radiometric fingerprinting chain seamlessly integrated within a single low-cost SoC (Paper VI).

Robust Fingerprinting in Challenging Wireless Environments
We study the severe impact of the wireless channel on radiometric fingerprinting and present mitigation strategies to make the fingerprinting system more robust even in challenging environments. Based on the fundamental insight of how multipath propagation parameters impact the features in a particular and non-random way, we employ a data augmentation method based on structured channel simulation to optimize the fingerprint classifier strategically. On top of that, noise compensation and a feature denoising filter are used to enhance the system stability in noisy conditions with weak signals. We demonstrate the system performance with significantly improved robustness across various real-world environments (Paper IV). In another work, we concentrate on channel characteristics. We use the received signal strength as an indicator to monitor the channel behavior between transmitters and receivers. Based on the pattern of channel dynamics, we propose a spoofing detection system to augment wearable device security under body motions (Paper III).

A major portion of this dissertation tackles the challenges associated with low-power and secure wireless communication in practical deployment. Backscatter transmitters strongly depend on external emitter infrastructures. Our first contribution offers a low-power design, enabling a long communication range with reduced dependence on emitters. Radiometric fingerprinting encounters challenges in practical scenarios due to diverse wireless channels and costly system deployment. Our second and third contributions address these challenges, enabling low-cost deployment while providing a robust fingerprinting service.

1.3 Methods
This dissertation follows system research methodology to understand how systems function and interact. System research is crucial for deepening our understanding of complex systems and integrating technologies to tackle challenges in a dynamic, interconnected world. All my work involves the use of real hardware to study phenomena and evaluate solutions. We use quantitative research methods to analyze and assess system performance. In addition, Monte Carlo
and numerical simulations are used to decompose complex systems, model their components, and study their interactions in a controlled environment.

While the objectives of our experiments vary slightly, they can primarily be categorized into two types. First, experiments are used to quantify and characterize specific phenomena. For instance, in Paper I, we quantify the tunnel diode oscillator stability by conducting controlled experiments. More commonly, experiments are deployed to demonstrate and evaluate the system performance. For example, in Paper IV, we conduct experiments in different wireless environments to demonstrate the robustness of the designed system.

My dissertation centers around wireless communication. The inherent complexities and dynamics of wireless environments make experimental setups, measurements, and analysis necessary but difficult. We tackle this challenge in different steps to approach realistic scenarios. When the wireless environment is not the primary research focus, we perform experiments under ideal conditions. Small-scale experiments are carried out in offices with cables, completely avoiding the effects of multipath and interference. If the setup with cables is not feasible, we place devices at close yet sufficient distances, mitigating major impacts from surrounding wireless environments. Most large-scale experiments take place in an anechoic chamber isolated from complex wireless conditions. These experiments help us to set up a benchmark, such as the benchmark dataset in fingerprinting work. When the wireless environment is a critical impact factor, we perform experiments in a variety of everyday settings inside office buildings. Moreover, we also conduct outdoor experiments for long-range communication setups.

Due to the complexities of wireless systems and their interactions with the environment, developing, testing, and optimizing a system design solely through real-world experimentation is often challenging or impractical. Wireless systems can be complex with numerous variables at play. Simulations allow us to break down these complexities, model different components, and observe their interactions in a controlled setting. Monte Carlo simulation based on the statistical wireless channel is used in Paper IV to provide a systematic overview of the wireless channel’s impact. Paper V employs numerical simulations to model the variation of backscatter fingerprints under varying voltages, assessing the robustness of the fingerprinting technique. In this dissertation, all simulations are conducted in Matlab in a hybrid manner, with part of the input from actual measurements.

1.4 Roadmap

This dissertation is structured into two parts. Part I provides a comprehensive summary including six chapters, while Part II is a compilation of the research papers included in this dissertation. The remainder of Part I is organized as follows. Chapter 2 introduces background information to facilitate
understanding this dissertation. Chapter 3 describes a summary of the six papers that constitute the core of my dissertation. Chapter 4 discusses the development of low-power communication systems from commodity IoT radios, to state-of-art backscatter technologies, to beyond backscatter designs. Chapter 5 reviews existing research on radiometric fingerprinting that aligns with or complements my work. Chapter 6 expands the vision towards future directions and concludes the dissertation.
2. Background

This chapter establishes the essential background knowledge. My dissertation centers around wireless communication systems. This chapter begins with an introduction to the architecture of transmitters. Then we discuss the imperfections inherent in transmitter hardware. Lastly, we present the characteristics and dynamics of wireless communication channels.

2.1 Resource-constrained Embedded Systems

The fundamental building block of IoT is the integration of a wireless communication interface with an embedded system. Such a system commonly combines sensing, computing, and wireless communication capabilities to monitor, gather, and process data from the environment and transmit it to other devices.

![Diagram of embedded system components](image)

*Figure 2.1. In embedded systems, the sensing, computing, communication and security tasks are handled by hardware components sharing limited computation, memory and energy resources.*

The tasks of sensing, computing and communication are handled by specific hardware components. Sensors are responsible for converting the physical phenomena of interest into digital signal formats suitable for processing. The computing tasks are usually handled by an microcontroller unit (MCU) which performs local processing and orchestrates coordination among the hardware components. Within the scope of this dissertation, communication is managed by a radio frequency (RF) transceiver, which encompasses both digital and analog components.
In an era of rapidly expanding device interconnectivity, the emphasis on security to ensure confidentiality, integrity, and availability has become essential. In some cases, specialized security components, such as cryptographic co-processors and accelerators, or specialized designs, such as trusted execution environments, are employed to provide enhanced protection against potential threats.

All these hardware components share available computation, memory and typically energy resources. Advances in semiconductor and packaging technologies have now enabled diverse low-cost and low-power sensors operating on merely tens of microwatts power. Additionally, computing tasks, facilitated by constrained and low-power microcontrollers, require only hundreds of microwatts power \[27, 28\]. In contrast, the communication task implemented using commercial radio technologies, including BLE, ZigBee, WiFi, LoRa and Sigfox, still demands a few to tens of milliwatts power (see from the comparison in Chapter 3) \[5, 28\].

Embedded systems often face resource-related constraints, due to requirements concerning size, cost and battery lifetime. Depending on the specific application, various tasks may occupy varying proportions of the available resources. Addressing these constraints while maintaining the system’s functionality stands as a fundamental challenge in the design of embedded systems.

2.2 Conventional Radio Frequency Transmitters

In this section, we review the conventional RF transmitter architecture commonly employed in wireless communication. As illustrated in Figure 1.2, a transmitter, at an abstract level, comprises three primary components: the protocol stack, modulator, and RF stage. Different architectures are adopted for the implementation of the digital wireless transmitter. We start with the straightforward direct-conversion transmitter architecture. As shown in Figure 2.2, a digital computation unit follows the protocol stack performing encoding, data packaging and modulation tasks. Consequently, it generates an information-bearing baseband signal at a low frequency. After that, the baseband signal is mixed with a high-frequency carrier signal generated by a local oscillator (LO), and is directly up-converted to the desired RF transmission band. After up-conversion, the signal passes through an RF amplifier, increasing its power to the desired transmission level. Typically, a band-pass filter is added after the amplifier to eliminate any unwanted harmonics and spurious emissions. The resulting signal is then transformed into an electromagnetic wave and radiated out through an antenna. The electromagnetic wave propagates at the speed of light and can be received by a receiver in its communication range.

As shown in Figure 2.3, the heterodyne architecture introduces an additional frequency conversion stage, where the baseband signal is first converted
Figure 2.2. The architecture of direct-conversion transmitters.

to intermediate frequency (IF) band for processing, then mixed with a high-frequency LO converting to desired RF frequency. The superheterodyne transmitter is an advanced version of the heterodyne transmitter, typically involving multiple conversion and amplification stages to produce high-quality RF signal for transmission. Unlike in the direct-conversion transmitter, the filtering and amplification take place at low frequencies during IF stages. Low-frequency filters and amplifiers are more stable, with higher gain and reduced power draw. By selecting the well-tuned IF and filters, the transmitter can attenuate image frequency signals, minimize phase noise, and counteract direct current (DC) offset.

Figure 2.3. An example of heterodyne transmitters.

The direct-conversion transmitter stands out for its simplicity and compactness due to the elimination of IF stages. This architecture inherently leans towards a low power draw. With the advancements in integrated circuit (IC) design and digital signal processing (DSP), direct-conversion designs have become more common, especially in modern digital communication systems where integration and miniaturization are priorities. On the other hand, while heterodyne or superheterodyne architectures are more complex, they provide enhanced selectivity and flexibility. These architectures are widely used in commercially off the shelf (COTS) transceiver chips that demand high performance. Furthermore, their inherent flexibility makes them adept at supporting a wide variety of wireless standards and modulation schemes. This versatility renders them ideal for multimode, multiband signaling schemes [29].
2.3 Backscatter Transmitters

Backscatter tags operate by reflecting and absorbing an external RF signal to convey information. The incident RF signal can be a modulated ambient signal, such as a TV, WiFi or Bluetooth signal, or an unmodulated carrier. We consider an unmodulated carrier in this dissertation, noted as \( S_{in}(t) = Ae^{j(2\pi f_c t + \phi)} \) where \( A \) is the amplitude, \( f_c \) is the carrier frequency and \( \phi \) is the phase of the incoming signal. This unmodulated carrier signal is generated by an external emitter device. The fundamental principle behind reflecting the RF signal is the impedance mismatching between the antenna complex impedance \( Z_A \) and the load complex impedance \( Z_L \). The resulting complex reflection coefficient is given as:

\[
\Gamma = \frac{Z_L - Z_A}{Z_L + Z_A} \tag{2.1}
\]

This coefficient describes the ratio of the incoming signal \( S_{in}(t) \) and the reflected signal \( S_{out}(t) \) as below:

\[
S_{out}(t) = \Gamma S_{in}(t) = |\Gamma|Ae^{j(2\pi f_c t + \phi + \angle \Gamma)} \tag{2.2}
\]

The process of modulating data onto the carrier involves changing the load impedance \( Z_L \), typically achieved by switching between different impedance configurations using an RF switch. For example, consider the three special cases: \( Z_L = Z_A \), \( Z_L = \infty \), and \( Z_L = 0 \). In these cases, \( \Gamma \) equals 0, 1, and \( -1 \), respectively, and the incident signal is reflected with corresponding amplitude and phase shifts. In the simplest case, \( \gamma \) is either 0 or 1, representing either fully absorbing or reflecting the incident signal. The fundamental idea of backscatter is switching between more than two load impedances to obtain the reflection and absorption states.

![Figure 2.4. The architecture of backscatter communication systems.](image)

A digital backscatter tag consists of two major modules, i.e., the control unit and the RF frontend. We show a 2-FSK backscatter system in Figure 2.4. The two impedances \( Z_L = 0, Z_L = \infty \) are selected. A low-power MCU or field programmable gate array (FPGA) serves as the control unit, generating a control signal embedding information on two frequency tones. This control signal toggles the RF switch at different speeds, inducing impedance changes to reflect and thereby modulate the incident RF carrier. By selecting specific
impedances and switching between them, the circuit can play different roles in the transmission process, such as shifting the phase or/amplitude in modulation schemes [12, 30, 31, 32, 33, 34] or suppressing the mirror image in single sideband backscatter [8, 14].

**Backscatter Harmonics.** Backscatter tags achieve modulation by toggling RF switches controlled by information-bearing signals. As a result, the signals are mixed with the incident carrier. Given the RF switches are digital components, the control signal is inherently limited to a square waveform. During the mixing process, the square signal introduces undesired harmonics. A square wave can be written as a Fourier series, which is the summation of sinusoidal components located at the harmonics of its fundamental frequencies:

\[
square(t) = \frac{4}{\pi} \sum_{n=1}^{\infty} \frac{1}{n} \sin(n \cdot 2\pi ft), \quad \text{where } n = 1, 3, 5, \ldots \tag{2.3}
\]

**Backscatter System Communication Range.** In backscatter systems, the received signal strength is determined by multiple parameters, including the carrier emitter transmission power \(P_t\), the distance between the emitter and the tag \(d_1\), the distance between the tag and the receiver \(d_2\), the wavelength of the RF signal \(\lambda\), the reflection coefficient of the tag \(\Gamma\), the loss in energy due to modulation \(\alpha\), and antenna gain \(G_t, G_{tag}, G_r\). The signal strength at the receiver can be modeled with Friis path loss model [9]:

\[
Pr = \left( \frac{P_t G_t \lambda^2}{(4\pi d_1)^2} \right) \left( \frac{G_{tag}^2 \alpha^2 \Gamma^2}{4} \right) \left( \frac{G_r \lambda^2}{(4\pi d_2)^2} \right) \propto \frac{1}{(d_1 d_2)^2} \tag{2.4}
\]

There are three key terms. The first term models the signal generated by the emitter and then propagates to the tag. Similarly, the third term models the signal propagation from the tag to the receiver. The middle term models the backscatter reflection operation. Given a backscatter system with fixed configuration \((P_t, \lambda, \alpha, \Gamma, G_t, G_{tag}, G_r)\) and receiver sensitivity (the minimum \(P_t\) that a receiver can detect), the maximum product of \(d_1\) and \(d_2\) is determined. The communication range \(d_2\) is large when the emitter is located close to the tag (small \(d_1\)). Conversely, as the distance \(d_1\) to the emitter increases, the communication range of the tag \(d_2\) diminishes.

### 2.4 Tunnel Diode

Tunnel diodes, a type of semiconductor diode, are notable for their ability to operate at very high frequencies. They were discovered more than half a century ago [35] and were the first semiconductor devices to demonstrate quantum tunneling. Quantum tunneling is a physics phenomenon that allows electrons to move through a barrier (P-N junction in the diode) even when they lack the energy that would normally be needed to do so.
Tunnel diodes have several remarkable properties. The most distinctive feature is its region of negative resistance (RNR) [36]. Consider biasing a tunnel diode with a DC voltage and varying the bias voltage ($V$). This relationship between the current ($I$) passing through a tunnel diode and the bias voltage ($V$) is nonlinear. When plotted on a current-voltage ($I$-$V$) graph, the curve of the tunnel diode first rises and then dips down, indicating negative differential resistance. Moreover, tunnel diodes operate at low voltages. In Figure 2.5, we display the RNR for the GE 1N3712 tunnel diode [37] used in Paper I and Paper II. It only consumes tens of microwatts of power to bias the tunnel diode to the RNR. Additionally, tunnel diodes can switch on and off very rapidly, making them useful in high-frequency applications. Tunnel diodes can be used in the design of logic elements, switching circuits, and RF components, such as amplifiers [36] and oscillators demonstrated in our work.

2.5 Imperfections in Wireless Transmitters

In the production of electronic components, inherent variabilities invariably introduce non-ideal characteristics, often termed as imperfections. This is due to limitations in manufacturing techniques. As a result of a compromise between cost and performance, such imperfections are considered acceptable. In a wireless transmit chain, the imperfections present in both analog and digital components [38, 39]. Hardware imperfections manifest themselves in the radiated signal through deviation from the standardized ideal signal. Such deviation stems from the individual components but is also affected by the adopted transmitter architecture.

For analog components such as amplifiers, mixers and filters, non-linear behavior and phase noise lead to distortion and unwanted harmonics. Oscillators bring along a unique set of challenges, including frequency offset, instability, spurious oscillation, phase noise, harmonic distortion, and amplitude instability. Passive components, such as capacitors, resistors, and inductors, are manufactured with particular tolerance levels, meaning that their actual values might differ from the stated ones by a certain margin. For quadrature transmitters, any imbalance between the in-phase and quadrature signals can lead to distortion in the modulated symbols. The digital-to-analog con-
verter (DAC) components, which bridge the digital and analog segments of signal processing, possess a fixed resolution, limiting the accuracy of signal depiction. Moreover, their non-linearity can further induce further harmonic distortion in the output signal. While digital components are manufactured to a more consistent standard, they often rely on clock signals produced by oscillators. The impairments of oscillators can introduce imperfections such as jitter and skew in the time domain. Furthermore, both analog and digital components emit thermal noise, which can degrade the quality of the transmitted signal.

2.6 Radiometric Fingerprinting Systems

Radiometric fingerprint systems are designed to identify wireless transmitters based on the unique characteristics of their transmitted signals. This concept was initially investigated in the radar community for military applications, including spectrum management, traffic analysis, or targeted transmitter tracking [40]. A radiometric fingerprinting system consists of multiple wireless transmitters and a fingerprint identifier or authenticator on the receiver side.

![Radiometric Fingerprinting System](image)

Figure 2.6. The overview of a radiometric fingerprinting system for identification.

Radiometric fingerprinting systems operate in three stages: signal acquisition, fingerprint feature extraction, and device identification or authentication through classification. Transmitters generate RF signals for communication following a specific physical-layer standard. As shown in Figure 2.6, once signals are acquired, their characteristics are extracted as distinct features, forming fingerprints. Subsequently, a matching algorithm determines the identity of the extracted fingerprint, drawing on knowledge from previously enrolled profiles. For an authentication service, the receiver executes the demodulation and obtains the transmitter’s self-claimed identity decoded in the data frame, such as the media access control address (MAC) or Internet protocol (IP) address. By checking the alignment between self-claimed identity and fingerprint identity, the system makes a decision whether to accept or reject the authentication request.

The fingerprinting system fundamentally tackles a classic pattern-matching challenge, spanning both enrollment and testing phases. When a device initially joins the network, its fingerprint is recorded at the receiver and registered as a device profile associated with its identity. Profiles for all enrolled devices are stored in a library, which aids decision-making in the testing
phase. To implement pattern matching, there are mainly two types of algorithms. Traditional methods measure the difference between captured fingerprints and enrolled profiles using distance metrics in the feature space, such as the Euclidean distance or Mahalanobis distance. The state-of-art methods utilize machine learning (ML) algorithms, such as support vector machine (SVM) [22, 41, 42, 43], k nearest neighbor (kNN) [22, 43, 44], or neural network (NN) [45]. These ML algorithms are trained with enrolled profiles. During the testing stage, the well-trained model judges the identity linked to the corresponding fingerprint. ML algorithms exhibit advantages in exploring nonlinear properties and handling outliers and obviate the need for the tedious distance threshold presetting process of the traditional methods.

2.7 Machine Learning Classifiers

This dissertation employs multiple supervised ML algorithms in different system designs. The basic supervised learning problem can be summarized as finding an accurate approximation, $f^*$, of a function, $f$, that expresses the relationship between the target variable $Y$ and the input variable $X$:

$$f : X \rightarrow Y$$

(2.5)

Supervised learning uses a finite labeled training set to approximate the function $f$. As the training set is fed into the algorithm, the algorithm $f^*$ adjusts its parameters until it has been fitted appropriately and outputs a well-trained model. In the classification problem, $Y$ is discrete class $C_k$ where $k = 1, \ldots, K$. Classification aims at taking an input vector $X$ and assigning it to one of the $C_k$ classes [46]. In this dissertation, four supervised learning classification algorithms were used in the fingerprinting system evaluation, including SVM, kNN, random forest (RFC) and NN.

**SVM** operates by finding optimum hyperplanes that separate the data points into classes. These hyperplanes are determined by support vectors, which are data points near the hyperplane. The optimum hyperplane maximizes the margin from these points. To prevent overfitting, a soft margin is used to allow some misclassifications for better generalization. For non-linear data, SVM uses kernel tricks to project data into higher dimensions. Commonly used kernels are polynomial, radial basis function (rbf) and sigmoid. While primarily for binary classification, SVM can extend to handle multi-class classification problems using one vs rest (OvR) or one vs one (OvO) strategies. In testing, the class with the highest confidence from all SVMs is selected as output.

**kNN** is a machine learning algorithm that uses all training data points directly. For classification, it identifies $k$ closest training samples to a new instance based on distance metrics like Euclidean or Manhattan distance. The class is then predicted using a majority vote from these neighbors. The choice of $k$
is crucial. A small $k$ can lead to an overfitting problem and the result only captures noise with high variance. However, a large $k$ can oversimplify the model, causing high bias and reduced accuracy.

**RFC** is an ensemble learning method that combines multiple decision trees to produce a more accurate and stable prediction. By using random subsets of data and features for each tree, it mitigates overfitting and offers robust performance. A decision tree is a supervised ML algorithm used for both classification and regression tasks. It works by splitting the data into subsets based on feature values, making decisions at each node until it reaches a leaf node with a predicted outcome. The tree structure consists of root nodes, internal nodes, and leaf nodes, representing decisions, conditions, and outcomes, respectively.

**NN** consists of interconnected neurons arranged into layers: an input layer, one or more hidden layers and an output layer. Each neuron applies a linear transformation by computing a weighted sum of its input, adds a bias, and then applies an activation function that introduces non-linearity. A loss function is used to measure the error between the predicted output of the network and the true target values in the training data. Training involves forward propagation of data and backward adjustment of weights using optimization techniques like gradient descent, based on the error from a loss function. For classification, the number of output layer nodes matches the number of classes.

### 2.8 Wireless Channels

Wireless communication channels are dynamic due to noise, interference, and unpredictable changes because of user movement and environmental shifts. Even minor environmental changes can significantly distort the received signal magnitude, phase, and frequency. In indoor settings, small-scale fading dominates, resulting from *multipath* propagation and *Doppler* shift. Additionally, the transmitted power dissipates and interference alters the signal strength and background noise, that is *signal to noise ratio (SNR)*.

![Figure 2.7. The model of a wireless communication system.](image)

The wireless signal, radiated out from the transmitter antenna, passes through the wireless channel in the air and reaches the receiver antenna. As shown in Figure 2.7, considering transmitter imperfections $f(t)$, the signal radiated out
from the transmitter \(s(t)\) can be modeled as:

\[ s(t) = i(t) * f(t) \]  

(2.6)

where \(\ast\) is the convolution operation and \(i(t)\) is an ideal information-bearing signal. Hardware-specific fingerprints deviate the transmitted signal from the ideal signal. When considering wireless channels \(h(t)\), the signal observable to the receiver \(r(t)\) can be modeled as:

\[ r(t) = s(t) * h(t) = i(t) * f(t) * h(t) \]  

(2.7)

The received signal involves both the transmitter fingerprint and the transient impact from the channel.

The multipath phenomenon occurs when a signal takes multiple paths from the transmitter to the receiver due to reflection, refraction, scattering, and diffraction off various obstacles and surfaces such as furniture, walls, glass and even the ground. This results in multiple signal versions at the receiver, differing in time, phase, and amplitude. In practice, a multipath channel has an inherent time-varying nature, which leads to fading. To model this, ray-tracing techniques are used, representing waveforms as particles and tracing their paths. Due to the radio channel’s variability, statistical models are often preferred, modeling each path with a random time-varying impulse response, statistically characterizing signal attenuation [47]. The channel equivalent time-varying impulse response of a single path can be written as below:

\[
    h(\tau,t) = \alpha(t)e^{-j\phi(t)}\sigma(t - \tau(t)) + n(t)
\]  

(2.8)

where the path delay \(\tau(t)\) is determined by the path length \(d(t)\) with Equation

\[
    \tau(t) = r(t)/c.
\]

\(\phi(t) = 2\pi f_c \tau(t) - \phi_D(t)\), where Doppler phase shift is captured by \(\phi_D(t)\), and Doppler frequency shift is \(f_D = \frac{1}{2\pi} \frac{d\phi_D(t)}{dt}\). Noise \(n(t)\) follows the normal distribution. \(\alpha(t)\) is a function of path loss. For a channel with \(N\) paths, the channel response \(h(t)\) is the sum of every path:

\[
    h(t) = \sum_{i=0}^{N} \alpha_i(t)e^{-j\phi_i(t)}\sigma(t - \tau_i(t)) + n(t)
\]  

(2.9)

Each path in the above multipath model can be modeled as a random process. The received signal in-phase \(r_I(t)\) and quadrature \(r_Q(t)\) components can be expressed with two Gaussian random variables with mean zero and equal variance \(\sigma^2\). The amplitude \(z(t) = |r(t)| = \sqrt{r_I(t)^2 + r_Q(t)^2}\) can be characterized by a Rayleigh distribution when there is no line of sight (LOS) path:

\[
    p_Z(z) = \frac{z}{\sigma^2}e^{-\frac{z^2}{2\sigma^2}}, \quad z \geq 0
\]  

(2.10)

If the channel has a fixed LOS component, the received signal equals the superposition of a complex Gaussian component and a LOS component. The
signal amplitude can be shown to have a Rician distribution given by:

\[ p_Z(z) = \frac{z}{\sigma^2} \exp\left[-\left(\frac{z^2 + s^2}{2\sigma^2}\right)\right] I_0\left(\frac{zs}{\sigma^2}\right), \quad z \geq 0 \] (2.11)

where \(2\sigma^2\) is the average power in the non line of sight (NLOS) multipath components and \(s^2 = \alpha_0^2\) is the power in the LOS component. \(I_0\) is the modified Bessel function of zeroth order.

The time variation of the channel due to the relative significant motions of either transmitter, receiver, or surrounding environment causes a Doppler shift in the received signal. The Doppler frequency shift is given by:

\[ f_D = \frac{v \cos(\theta)}{c} f_c \] (2.12)

where \(c\) is the velocity of signal propagation and the object is moving at a spatial angle \(\theta\) with velocity \(v\). The Doppler power spectrum \(S(f)\) represents how the power of the received signal is distributed across various frequency shifts due to the Doppler effect. Different propagation environments and mobility patterns result in various shapes for \(S(f)\). With uniformly distributed scatterers around a consistently moving receiver, \(S(f)\) is bell-shaped and zero beyond \(\text{max}(f_D)\).

The noise term \(n(t)\) in Equation 2.8 and 2.9, can be represented using additive white gaussian noise (AWGN). The noise exhibits a flat spectrum across frequencies and follows a Gaussian distribution in time. The AWGN model simulates the random noise that remains independent of and can be added to the signal of interest.
3. Summary of Papers

Paper I

DOI: 10.1145/3498361.3538923

Summary

In this paper, we introduce Judo, a low-power transmitter architecture based on a tunnel diode self-oscillating mixer (SoM). We refer to the conventional transmitter architecture and simplify the RF stage using a tunnel diode SoM circuit. Our design achieves low-power radio transmissions by integrating the stages of a radio transmitter into a single power-efficient step. In this step, a high-frequency carrier signal is generated locally and mixed with the analog baseband signal with a peak power draw around 48 μW. Judo transmitters make trade-offs to achieve significant power efficiency, which is countered by a number of design choices. First of all, Judo transmitters radiate weak signals. We employ a highly sensitive receiver to maintain a long communication range. Secondly, Judo transmitters trade stability for power efficiency. While Judo transmitters can operate independently, the communication quality is degraded in dynamic environments. To address this issue, we stabilize the transmitter using the injection-locking phenomenon. An emitter device generates a carrier signal to stabilize the transmitter. As Judo transmitters can latch onto even weak signals, it allows the emitter device to be located significantly far away from the transmitter, and still support a long communication range. Based on this architecture, we designed a transmitter with a frequency-shift keying modulation scheme. Our evaluation demonstrates that Judo transmitters can achieve a communication range exceeding 100 m, even when the transmitter is located at 100 m away from a carrier emitter generating a signal of 25 dBm strength.
Reflections
This paper was my initial attempt at exploring low-power transmitter design using tunnel diodes. This was also my first collaboration with Prof. Ambuj Varshney, who had done work enhancing backscatter communication through tunnel diodes. Initially, our focus was characterizing the instability of tunnel diode oscillator (TDO). In our experiments, we discovered a design that can feed baseband signals into the TDO, enabling the circuit to serve dual roles as a SoM. This discovery formed the basis of Judo. The main contribution of this work is a novel low-power transmitter architecture, using a design principle fundamentally distinct from long-existing backscatter. Judo architecture closely resembles a conventional transmitter but operates on a low power budget similar to a backscatter transmitter, without inheriting its limitations. We demonstrate a new way to achieve wireless communication with power consumption at the microwatt level. This research not only prepared me with solid knowledge about low-power transmitter design but also laid the groundwork for subsequent research on backscatter fingerprinting (Paper V).

My Contributions
In this paper, Prof. Ambuj Varshney and I share the co-primary authorship. This work builds on Ambuj’s earlier TDO research [48, 49]. Together, we discovered the SoM property. Based on this discovery, we designed the Judo transmitter architecture. We conducted experiments collaboratively. I took the lead in analyzing the experimental data, interpreting results, and processing graphs. I contributed to writing and structuring the manuscript.

Paper II

Summary
In this paper, we introduce a light receiver architecture allowing commodity IoT devices to receive light fidelity (LiFi) transmissions using their existing radio transceivers. We named this system TunnelLiFi. The key component is a TunnelLiFi tag acting as a LiFi-RF bridge that replicates information contained in LiFi transmissions onto radio waves. TunnelLiFi tag evolves from Judo transmitter architecture by coupling a photodiode with the tunnel diode
SoM. The SoM property allows the mixing of a weak photodiode signal with a locally generated RF carrier signal, drawing less than 100 μW of power. The system consists of a LiFi transmitter, the TunnelLiFi tag, and an RF receiver. An external LiFi transmitter, built on open VLC project [50], is used to modulate the light and transmit baseband information. The TunnelLiFi tag’s photodiode receives this signal and the tunnel SoM mixes it with an RF carrier signal. As a result, radio transmissions from the TunnelLiFi tag can be received by commodity radio transceivers. Similar to Judo, an external weak carrier signal stabilizes the emitted RF signal via injection locking, enhancing the link quality. Our prototype demonstrates TunnelLiFi as a low-power bridge that can operate with low bitrates (2.93 kbps) even in low-light conditions (300 lux).

Reflections
This paper is a follow-up work of Judo, and considers a novel application of tunnel diode SoM, acting as a LiFi-RF bridge. LiFi simultaneously provides energy, illumination, and information, making it a good candidate for IoT connectivity. However, adopting LiFi technology requires changes to both the transmission and reception infrastructure. This work focuses on the receiver design challenge and rethinks the LiFi receiver architecture. TunnelLiFi design provides a low-power solution enabling pervasive radio transceivers to receive LiFi signals. The contribution of this work lies in its potential to facilitate widespread deployment of the LiFi technology.

My Contributions
Muhammad Sarmad Mir and I share the co-primary authorship for this work. Prof. Ambuj Varshney proposed the idea of leveraging a Judo transmitter architecture to design a LiFi-RF bridge. Together, we performed the proof-of-concept experiments. Sarmad then conducted subsequent experiments during his visit to NUS (University of Singapore). I analyzed the experimental data and also contributed to writing the manuscript.

Paper III
Summary
In this paper, we introduce a spoofing intrusion detection system (IDS) for wearable devices leveraging body motion and lightweight statistical algorithms. This system utilizes the received signal strength indicator (RSSI) time series to monitor channel behavior between the transmitter and receiver. With body movement, the RSSI time series pattern of off-body devices differs from the on-body devices, and the on-body devices at different positions also differ from each other. This diversity in channel dynamics is used to identify frames that violate the regular pattern of the wireless signal from legitimate wearables. The system is evaluated using bodyworn devices positioned differently in real-world attack scenarios, including both on-body and off-body attackers. The results show that the system can detect naive attackers accurately, and maintains good accuracy even for sophisticated attackers.

Reflections
This short paper is the first project at the beginning of my Ph.D. journey, marking the start of my exploration in the wireless mobile computing research field. I am funded by a project focusing on securing wearables and implants. The initial idea is to leverage the wireless channel dynamics due to human body movement to secure the system. RSSI, a common link metric easily collected via transceiver chip APIs, offers low-resolution information about the transmitter’s location. While RSSI has been used in stationary setups to detect spoofing devices, our work investigates its dynamics in wearable contexts. We find that RSSI time series exhibit distinct patterns during body movements, which are leveraged to identify the transmitter in this project.

My Contributions
I am the lead author of this paper. I identified and formulated the research question with feedback from Prof. Christian Rohner. I designed the framework and developed the algorithms. Sam Hylamia set up the wearable testbed. I implemented the system and conducted all experiments. I wrote the manuscript with inputs from all co-authors.

Paper IV
Summary

In this paper, we present RRF, a robust radiometric fingerprinting system designed to offer reliable identification/authentication under channel fading disturbances. Complex wireless channels significantly impact the accuracy of identifying transmitter devices using radiometric fingerprints. Through practical experiments and systematic simulations, we decompose the channel impact factors into three components, multipath, noise, and Doppler. We analyze the impact of each factor. The results indicate that multipath and noise dominate the distortions, and affect fingerprint features in a specific way. Leveraging these insights, we design pertinent strategies to deal with different channel impacts. Our system deploys a hybrid pipeline that combines wireless channel simulation, signal processing, and machine learning. In this pipeline, RRF first utilizes structured channel simulations to adjust the decision boundaries of the fingerprint classifier, which improves its tolerance towards multipath channel interference. Then, in the identification phase, RRF relies on noise compensation and a feature denoising filter to augment the system’s stability for weak reception signals. Our experimental results show that RRF achieves an average accuracy consistently above 99% in empirical scenarios with complex channels.

Reflections

In the prior PHY-IDS project, I learned about the limitations of channel characteristics for transmitter identification, such as the ease of spoofing and sensitivity to device movement. In pursuit of a more effective alternative, I turned to radiometric fingerprinting and followed the work of Brik et al. [22]. Although not a new concept, my experimental results reveal the significant impact of wireless channel interference on radiometric fingerprinting. Most existing research emphasizes the ability to differentiate devices and often omits necessary designs to enhance the robustness under diverse wireless channel conditions. To bridge this research gap, I started the RRF project. This work contributes to understanding the impacts of wireless channels on radiometric fingerprints. We suggest using pertinent strategies to deal with different channel impacts. We demonstrate that already a simple ML model, combined with an established feature representation and modest signal processing, is sufficient to take the robustness of the radiometric fingerprinting system to a new level.

My Contributions

I am the primary author of this work and the sole student contributor. I identified and formulated the research question. I proposed the solution employing simulation-based data augmentation to enhance the system’s robustness.
Then, the idea is further completed together with Prof. Christian Rohner. I designed the system framework and implemented it. I conducted all the experiments. I also wrote the majority of the manuscript.

Paper V

Summary
In this paper, we present a backscatter radiometric fingerprinting system designed to identify tags and carrier emitters. The passive radiometric fingerprinting approach fits well with backscatter systems, which prioritize low power and simplicity. Backscatter systems delegate the power-intensive generation of high-frequency carriers to an external emitter device, while the low-power tag modulates data by reflecting these carrier signals. In this paper, we systematically analyze the backscatter architecture and decompose the fingerprint. This allows us to accurately distinguish and classify both tags and carrier emitters with a true accept ratio of over 98.4% and below 1.6% false accept ratio. Fingerprinting backscatter systems are inherently more challenging than conventional radio transmitters due to the emitter-tag separation with a simple architecture. To offer a comprehensive perspective, we assess the importance of features in conjunction with three sets of conventional transmitters. In addition, we seek insights into fingerprint stability. Our finding suggests that tag fingerprints are susceptible to voltage variations.

Reflections
In this paper, I combined the backscatter knowledge and experience collected in the fingerprinting project. Existing backscatter fingerprinting works predominantly focus on radio frequency identification (RFID) systems. In RFID system, the carrier emitter is integrated with the receiver, which overlooks the diversity of carrier emitter devices. The major contribution of this paper is demonstrating the feasibility of fingerprinting both the tag and the emitter device. Unlike fingerprinting commercial off-the-shelf (COTS) transmitters, collecting a large number of backscatter tags is challenging due to the lag in commercialization. In the evaluation, we used ten tags based on Carlos Pérez-Penichet’s design.
My Contributions
I am the lead author of this paper. I identified and formulated the research question. I designed the framework and implemented the system. The data collection was a collaborative effort with co-authors. I conducted all evaluation experiments. I wrote the manuscript with feedback from all co-authors.

Paper VI

Summary
In this paper, we demonstrate the first fully integrated on-board radiometric fingerprinting system, designed for low-cost and low-power COTS receivers. While radiometric fingerprinting systems have proven effective, existing systems require specialized hardware and non-trivial computational capability to extract fingerprint features, hindering their widespread adoption. To advance towards practical deployment, we transfer the entire fingerprinting pipeline, including signal acquisition, feature extraction and classification, to an nRF52833 embedded SoC that costs under 6 dollars. We achieve raw signal acquisition at a high sampling rate by re-purposing the Bluetooth direction finding extension (DFE) functionality. Our fingerprint feature extraction builds on a lightweight coherent-receiver pipeline, and we deploy a simple classification model for identity determination. The final prototype can identify a single frame within one second, consuming energy equivalent to half a second of active channel monitoring. Based on the experiments with 32 transmitter devices, our system consistently delivers over 92% identification accuracy.

Reflections
In this paper, we address another practical challenge in radiometric fingerprinting. In our earlier endeavors in fingerprinting, we were disappointed by the reliance on cumbersome and expensive setups requiring costly software defined radio (SDR) and dedicated computer processing resources. This project began with a critical question thrown out by Prof. Christian Rohner about the feasibility of a cost-effective implementation. The bottleneck to achieving it is the acquisition of signals at appropriate sampling rates using low-cost receivers. Typically, embedded transceiver chips offer only limited statistical link quality metrics rather than providing access to raw signal data. However,
the emerging support for the Bluetooth DFE in COTS devices changes this situation, paving the way for this project. For the first time, ORF showcases a radiometric fingerprinting system fully integrated on a cheap and low-power SoC.

My Contributions
I am one of the lead authors sharing co-primary authorship with Mikolai-Alexander Gütschow, a Master’s thesis student under my supervision. Prof. Christian Rohner and I formulated the research question. I suggested using the nRF52833 SoC and investigated the viability of implementing our system on it. Mikolai adapted the system architecture from the RRF project for embedded implementation by translating it effectively into C code. We conducted the evaluation experiments together. I wrote the manuscript with feedback from all co-authors.
4. From Commodity IoT Radios to Backscatter to Beyond Backscatter

This chapter presents an overview of IoT communication radios that are commercially available and backscatter radios described in related research. By offering a horizontal comparison across different radios, this chapter contextualizes the results presented in Paper I within a broader context.

4.1 Commodity IoT Radios

In the realm of IoT, a variety of commercial radios, i.e. WiFi, Bluetooth, ZigBee/Thread, Sigfox, and long-range radio (LoRa) are employed to meet the varying demands of data rate, communication range, and power consumption, thereby ensuring optimal performance across diverse applications and deployment scenarios. Each radio technology has carved out its own niche in the marketplace. WiFi, with its high data rates reaching up to 9.6 Gbps in WiFi 6, is preferred in scenarios demanding transmission of large data volumes with low latency. However, it is characterized by relatively high power consumption. In contrast, Zigbee/Thread and BLE (Bluetooth Low Energy) are optimized for low power consumption. BLE is suitable for short-range point-to-point communications with up to 2 Mbps data rate [51]. Zigbee/Thread, with IEEE 802.15.4 as the underlying physical and network layer protocol, is optimized for reliable and low-power mesh networks with lower data rates of 250 kbps [52]. For long-range, low-power IoT applications, LoRa and Sigfox are the preferred protocols. LoRa, utilizing chirp spread spectrum (CSS) modulation, can achieve a communication range of several kilometers, with low data rates ranging from 0.3-50 kbps [53]. Sigfox, while operating at even lower data rates of up to 100 kbps, maintains a comparable range and is similarly efficient in terms of power usage [54].

In general, commercial radios have relatively high power consumption, typically consuming a few to hundreds of milliwatts. Appendix A lists parameters determining the energy efficiency of the aforementioned technologies, collected from the datasheet of COTS transceiver chips or SoCs widely used in IoT applications. In the following section, Figure 4.1 compares Judo with commodity and backscatter radios to provide a holistic view of the dissertation contribution in a broader context.
4.2 Backscatter Radios

RFID and Modified RFID for Sensing. RFID remains the dominant backscatter application in the market. It is known for its battery-free design with extremely low power consumption and is primarily used for short-range identification and tracking. As the IoT landscape expands, RFID sensing integration has gained traction, with manufacturers like NXP [55], Smartrac [56], Radio-Force [57] offering RFID sensors for environmental monitoring. Open-source platforms like WISP [58] and Moo [59] enable advanced low-power applications, such as battery-free camera [60] and microphones [61]. However, these systems rely on complex, costly, and power-hungry RFID readers which can only interact with RFID tags.

Backscatter Beyond RFID. Extensive studies enhance the backscatter system communication performance and compatibility with existing commodity radios, which unlock the potential of backscatter beyond RFID. Some works deploy unmodulated excitation signals like RFID and enable the interoperation with commodity radios, such as WiFi [14], Bluetooth [17, 62], IEEE 802.15.4 [8]. Passive WiFi demonstrates tens of meters communication range with up to a few Mbps data rate [9]. LoRa backscatter presents a CSS backscatter design, enabling long-range communication at a few kilometers [12]. LoRea focuses on narrow-band communication, demonstrating a comparable communication up to kilometers [11].

Ambient Backscatter Ambient backscatter modulates information onto RF signals already present in existing systems. These ambient signals often carry their own data. Early attempts, using TV signals [63] and WiFi transmissions [64] for tag communication, face range (below a few tens of centimeters), and data rate (a few kbps) limitations. BackFi [65] improves the data rate to Mbps levels but is still constrained to a short communication range of a few meters. FS-Backscatter mitigates the carrier signal interference by shifting the excitation signal to an adjacent spectrum [66]. HitchHike increases the communication range to tens of meters with codeword translation [15]. Syncscatter further extends the communication range of ambient backscatter by symbol-level synchronization [10]. PLoRa utilizes ambient LoRa transmission, enabling long-range backscatter communication [13].

Energy per Bit Comparison. To assess how efficiently different communication technologies utilize the energy, the data rate is a key factor. Higher data rates may consume more power but can be more energy-efficient over time, making energy per bit a valuable metric. Figure 4.1 shows the active power draw in relation to data rates. The dotted diagonal lines represent energy per bit, indicating the amount of energy consumed to transmit a single bit of information. As illustrated, backscatter systems are typically around 100 times more efficient than conventional radios, which is advantageous for battery-constrained or energy-harvesting applications. When comparing backscatter systems described in related research, the power consumption largely depends
Figure 4.1. The energy per bit of Judo and wireless radio technologies, including commercial radios and backscatter systems introduced in research papers. The full dataset can be found in Appendix A Table 1 and 2.

on the implementation. An implementation using off-the-shelf discrete components can consume a magnitude higher power than the implementation in a sub-micron ASIC design. Most backscatter systems report below sub-hundred microwatts power consumption based on simulations. SyncScatter, being an exception, shows a 30 μW consumption in an IC prototype. Due to the absence of an accurate tunnel diode model, in Judo (Paper I), we estimate the power draw by measurement using a prototype with discrete components. While Judo is not a backscatter design, its energy per bit metric based on the prototype is comparable with existing backscatter works. In our work, we also demonstrate the potential to support higher data rates with even better efficiency.

Communication Range Comparison. In backscatter systems, where tags serve as the data source, the communication range of these tags is crucial. However, it is misleading to consider only the tag-to-receiver communication range. The maximum effective communication range in backscatter systems is determined by the product of emitter-to-tag ($d_1$) and tag-to-receiver ($d_2$), as described in Equation 2.4. Configurations with a very short emitter-to-tag
distance \( (d_1) \) are impractical for real-world deployments and fail to highlight the advantages of backscatter technology over conventional active radios. To effectively compare different backscatter systems, we use the range metric \( d_1d_2 \), considering both emitter-to-tag and tag-to-receiver distances. This metric is also applied to the Judo system, due to its similar bistatic configuration, ensuring a consistent comparison across systems. Figure 4.2 shows the communication range comparison of existing backscatter systems. The dotted diagonal lines show the \( d_1d_2 \) metric. The marker size represents the data rate. Among the listed systems, Judo outperforms most systems in range, except the LoRa backscatter system [12]. The remarkable communication range of LoRa backscatter is mainly attributed to the high link budget provided by CSS modulation and the uses of the sub-GHz frequency band. Judo operates at the same frequency band but deploys narrowband FSK modulation. Moreover, Judo uses 100 kbps data rate, which is at least three times higher than the data rate demonstrated in the LoRa backscatter. Notably, the design of Judo is modulation-agnostic and can also support CSS, offering the potential for further communication range enhancement.
4.3 Radio Technologies Beyond Backscatter

Backscatter systems divide tasks in the conventional radio RF stage into two devices. The emitter infrastructure manages the power-hungry task of generating high-frequency carriers. This allows tags to only perform low-power modulation operations by reflecting the incident carrier signal. While the power consumption of the backscatter tag is orders of magnitude less than conventional transmitters, the total energy required for every transmission across emitter, tag, and receiver devices is unchanged at best. Two approaches have emerged targeting an improvement in overall system energy efficiency, shifting beyond traditional backscatter.

**Reflection Amplifier.** One approach replaces the standard RF switch in a backscatter tag with a low-power reflection amplifier, allowing the carrier signal to be backscattered with a gain. Negative-resistance elements such as transistors or tunnel diodes are used in reflection amplifiers. Amato et al. create a tunnel diode-based reflection amplifier that operates in the 5.8 GHz band with a gain of 34 dB while consuming 45 μW [67, 68, 69, 70]. Adeyeye et al. design a repeater in the 5.8 GHz frequency band using a tunnel diode reflection amplifier with a gain as high as 50 dB while consuming 40 μW [71]. Varshney et al. design a tunnel diode reflection amplifier operating in the 868 MHz band with a reflection gain of 35 dB and a peak biasing power of 57 μW [48]. Kimionis design a transistor-based reflector in the 900 MHz frequency band with a reflection gain of 10.2 dB and a peak power consumption of 325 μW [72]. Dong et al. leverage tunnel diode reflection amplifier with 22 dB gain to strengthen weak GPS signals [73]. However, the reflection gain amplifies unwanted harmonics inherent in backscatter mixing operations as well, covering a significant part of the unlicensed spectrum. In Paper I and II, Judo transmitters are not reflection-based and leverage the SoM property to overcome this challenge.

**Modulated Noise Communication.** Another innovative approach is Modulated Noise Communication (MNC). This method moves one step further and directly eliminates the generation of high-frequency carriers, leveraging Johnson noise left by the transmitter resistors to modulate baseband data. Kapetanovic et al. demonstrate a prototype with a low data rate up to 26 bps and a limited range up to 7.3 m [74]. From the system view, MNC operates independently like a traditional transmitter, eliminating the complexity associated with the emitter in backscatter systems. Circuit-wise, MNC mirrors backscatter modulation by altering antenna loads, sharing the advantages of minimal power consumption and simplicity.
5. Transmitter Identification with Fingerprints in Wireless Signals

In wireless communication systems, fingerprints embedded in signals can be leveraged to identify transmitters. This chapter overviews existing fingerprinting works and puts the results presented in Paper III, IV, V and VI into a broad research field.

5.1 Channel-specific Fingerprints

Channel-specific fingerprints capture properties of the communication channel, which have gained popularity in indoor localization due to their simplicity and minimal hardware requirements. They leverage features such as received signal strength [75, 76, 77], channel state information (CSI) [78], and channel frequency response [79, 80]. Unlike conventional geometric localization approaches, fingerprinting methods employ pattern matching to determine device positions, which excels in cluttered indoor environments with stable configuration [76, 77, 78]. These systems commonly involve multiple measurements at different points in space to improve the precision and robustness of localization. Channel-specific fingerprints fit better in localization rather than device identification due to their low uniqueness. Using channel-specific fingerprints for device identification relies heavily on the assumption of stationary users and environments, making them hard to apply in practical scenarios where moving objects are pervasive in the environment. Paper III investigates wearable scenarios. Based on the pattern of body movement, channel-specific time series are utilized to identify transmitters.

5.2 Hardware-specific Fingerprints

Hardware-specific fingerprints highlight unique imperfections of electronics on the transmitter. We also term them radiometric fingerprints in this dissertation. Radiometric fingerprints were first investigated in the radar community as specific emitter identification (SEI) for military usage to track enemy radars during World War II [81]. Research emerged in the mid- and late-90s to detect illegal radio VHF FM transmitters [82, 83, 84]. With the proliferation of commodity wireless communication devices, radiometric fingerprints have been
explored to identify WiFi, ZigBee, Bluetooth, LoRa and RFID transmitters in diverse scenarios, including device cloning, defective device detection, and access control. In this section, I structure existing systems shown in related research into five categories.

**Transient-based Fingerprints.** Early works focus on the transient phase at the beginning of each radio transmission before frequency synthesizers settle on the pre-set frequency. Unique features of the transient signal, such as the amplitude profile [85], phase characteristics [86], adjacent Fourier transform spectra [23], and energy envelope [42] are leveraged for device identification. However, systems relying on the transient phase are sensitive to device locations and antenna polarization [23]. Moreover, the duration of the transient phase is usually at a sub-microsecond level, requiring high-speed signal acquisition equipment, which increases the implementation cost.

**Modeling-based Fingerprints.** A few works have focused on mathematically expressing hardware imperfections by modeling the signal transmission process in the transmit chain. These models primarily address nonlinear distortions from transmitter hardware, such as DACs and amplifiers [18, 25, 87]. The model parameters, obtained either from statistical modeling [87] or captured signals using data-driven regression [25], serve as device fingerprints. Modeling-based fingerprints commonly do not cover [18] or only partially cover the complex disturbance from wireless channels [18, 25].

**High-dimensional Raw Signal Fingerprints.** Neural network (NN) algorithms have been increasingly exploited to automate feature extraction in radiometric fingerprinting systems. The capability of NN allows the usage of high-dimensional signals in the time or frequency domain as raw fingerprints. Various formats of raw fingerprints are used, including I/Q signal time series [26], time series of error between ideal and synchronized signals [19], and constellation trace plots in image format [88, 89]. However, NN-based fingerprinting is significantly impacted by wireless channels. Al-Shawabka et al. quantify the impact with experimental data [20]. While raw signals contain all hardware-specific information, they are often very noisy. Leveraging NN to process raw signals demands extensive computation resources, making it less suitable for efficient fingerprinting. Additionally, NN are often criticized for their black-box nature, lacking in flexibility and interpretability. In our work, we adopt a white-box approach, focusing on analyzing and eliminating different channel disturbances.

**Low-dimensional Modulation Domain Fingerprints.** Modulation-based fingerprints assign statistics of the modulation errors as device fingerprints, which are extracted via well-defined signal processing procedures. Modulation-related features are first proposed by Brik et al. [22] for WiFi systems. The initial features include frequency error, SYNC correlation, I/Q offset, magnitude error, and phase error. Then this technique is explored in several ISM band technologies, including ZigBee [90], BLE [91] and LoRa [88, 92]. And features
are expanded to I/Q imbalance [41], modulation shapes [93], and constella-
tion error [94]. The fingerprints in this category are interpreted depending on
the underlying modulation scheme. However, with the guidance of wireless
communication knowledge, the crafted features refine structured fingerprints
containing more intense information. Modulation domain fingerprinting typ-
ically requires fewer resources than methods that process large volumes of
raw signals, making it efficient for embedded systems deployment with lim-
ited computation, memory and energy resources. This dissertation (Paper IV,
V, VI) focus on modulation domain fingerprints. We utilize the side infor-
mation from the demodulation pipeline, allowing for seamless integration into
the receiver chain with minimal additional resource demands.

**Fingerprints in Backscatter Systems.** Existing backscatter fingerprinting
works are mainly conducted in the RFID context. Features in time and fre-
quency domains are used to fingerprint RFID tags [95, 96]. Some works focus
on fingerprints unique in backscatter designs. RCID leverages the reflection
coefficient of RFID tag circuit as the fingerprint [97]. Hu-Fu uses the coupling
features between two tags to identify tags, eliminating the wireless channel
disturbance but requiring two tags to operate simultaneously [98]. Eingerprint
leverages distinct energy storage capability of passive RFID tags to authen-
ticate them [99]. HarvestPrint uses frequency variations during the capacitor
discharge in low-power tag oscillators to fingerprint tags [100]. All these sys-
tems only identify backscatter tags. In contrast, our work (Paper V) finger-
prints both tags and emitters for the first time. We move the scope to general
backscatter fingerprinting systems beyond RFID.

### 5.3 Channel-robust Fingerprinting Systems

Effective fingerprinting systems should maintain robustness against positional
changes, environmental variations, and device mobility. The wireless chan-
nel is a major contributor to accuracy degradation in radiometric fingerprint-
ing [18, 19, 20, 21]. To enhance the robustness of the radiometric finger-
pinting system towards complex channel disturbances, multiple methods are
proposed.

Several works investigate frequency domain features robust to locations to
fingerprint WiFi devices [101, 102]. These features leverage the relative re-
lation between subcarriers, which are limited to wideband radios. One ap-
proach alleviates channel effects by restoring a less distorted version of the
transmitted signal with the channel estimation support [25, 103]. However,
this approach requires either a known reference signal or sophisticated algo-
Rithms for channel estimation. Another approach deploys a transfer learning
method and retrains the model during deployment [104]. Training models is
often resource-intensive and time-consuming, which is not always practical in
actual deployments. Several works focus on the low signal strength condi-
tion and devise a hybrid classifier by adjusting feature weights based on the received signal SNR level \([90, 105, 106]\). However, these systems overlook other impact factors, such as multipath.

To make fingerprinting systems robust to unseen complex channels, Soltani et al. leverage CNN together with data augmentation, which expands the training set with various channel-distorted fingerprints considering both multipath and noise impacts \([21]\). Similarly, our work (paper IV) adopts data augmentation methods with the support of wireless channel models. Differently, our method decomposes different channel impact factors. With detailed guidance that benefits from the structured feature space, we deploy distinct strategies towards multipath and noise, resulting in a more explainable and efficient system design.
6. Conclusions and Future Work

6.1 Conclusions

Wireless communication is a key enabler for the IoT, connecting billions of devices. The increasing prevalence of constrained embedded devices, operating on limited energy resources, requires developing low-power and secure wireless communication solutions. This dissertation addresses this need by designing low-power wireless transmitters and developing passive radiometric fingerprinting systems for robust and efficient transmitter identification.

Our work contributes to understanding the role of transmitter components and channel factors in wireless communication systems. We present a novel strategy for designing low-power transmitters by rearranging the functions of components. By analyzing the impacts of wireless channel factors, we improve the robustness of the fingerprinting system under complex channel distortions. And by understanding how components contribute to signal generation, we engineer the fingerprint features enabling the resource-efficient implementation on embedded devices.

Backscatter has long been considered a promising candidate for low-power wireless communication. It facilitates the deployment of embedded systems using small batteries, such as thin film or printed types, or battery-free options that harvest energy from environmental sources. However, in real-world deployments, the communication range of backscatter transmitters strongly depends on an external emitter infrastructure. For an extended communication range, backscatter requires the proximity of the transmitter tags to an emitter device that generates strong signals.

In this dissertation, we propose a transmitter design that rethinks the principle of backscatter design. Instead of offloading power-hungry tasks to an external emitter, as backscatter does, we integrate these tasks into the RF stage in a single low-power manner. While resembling conventional transmitters, this design distinguishes itself through its remarkable power efficiency, consuming only tens of microwatts power. This efficiency comes at the cost of stability, which we address by using an external carrier emitter to stabilize the signal. We rearrange the functions of transmitter components and redefine the role of the emitter device. Our transmitter can latch onto a weak carrier signal, significantly reducing its dependence on emitters. This allows the transmitter to communicate over a hundred meters, even when the emitter is more than a hundred meters away. This contribution provides a viable solution to the range limitations inherent in backscatter systems, potentially expanding their application in various real-world scenarios.
Radiometric fingerprinting identifies wireless transmitters based on their hardware components’ unique and inherent imperfections. Its passive nature adds a valuable security layer, enabling the identification of transmitters without requiring active cooperation from the devices being identified. This method protects wireless communication systems from various security threats, including device cloning, spoofing, and replay attacks. This dissertation adapts radiometric fingerprinting for real-world deployments, applying it to both conventional active radios and backscatter radios.

The dynamic nature of real-world wireless channels significantly challenges the robustness of radiometric fingerprinting. The key idea is to decompose the complex channel interference into individual impact factors and systematically assess each factor’s impact on radiometric fingerprints. These understandings enable us to develop tailored strategies that effectively mitigate wireless channel disturbances. The proposed system substantially enhances fingerprinting robustness, achieving consistent identification accuracy across complex wireless environments. This work makes radiometric fingerprinting effective in real-world deployments under diverse scenarios.

Most prior radiometric fingerprinting works rely on sophisticated signal acquisition equipment and dedicated computer processing resources. Our work demonstrates a resource-efficient radiometric fingerprinting chain. From signal acquisition to feature extraction and classification, all are seamlessly integrated within the confines of a single SoC (Paper VI). This contribution simplifies the traditionally complex setup, facilitating the large-scale deployment of fingerprinting systems on low-cost COTS embedded devices.

Backscatter transmitters fundamentally differ from conventional active transmitters regarding the electronic component composition within the transmit chain. In backscatter systems, the radio transmission is a joint effort between the backscatter transmitter and the carrier emitter. Our work demonstrates the dual identification of backscatter transmitters and carrier emitters in different scenarios. Beyond security applications, recognizing the emitter embeds a notion of locality, exposing fingerprinting usage in backscatter network management tasks such as coordinating emitters.

In general, this dissertation is anchored in the rapidly evolving landscape of enhancing connectivity among IoT devices. Our works contribute to understanding the roles of components in wireless communication systems and demonstrate the practical implications of these insights in real-world applications.
6.2 Future Work

As we look toward the future, it is intriguing to speculate on how the ideas and results from this dissertation can be leveraged and expanded upon in new and innovative ways. In this section, I will discuss several directions as potential areas for future development and exploration.

**Stand-alone Tunnel Diode SoM Transmitters** The tunnel diode SoM transmitter, proposed in Paper I, employs a design principle of trading stability for low-power consumption. Experimental results demonstrate the transmitter’s capability of communicating across several floors indoors without requiring a carrier emitter. However, the radiated signal is noisy, unstable, and can be affected by changes in the environment. In our work, the injection-locking phenomenon is used to improve the tunnel diode SoM stability, which requires an external emitter. Like a backscatter system, employing an additional emitter device increases the complexity of system deployment. An open challenge is alternative methods to stabilize the tunnel diode SoM without increasing system complexity. The possible direction is leveraging another communication channel to feed the injection-locking signal, e.g., flexible conductive substrates [107].

**Fingerprinting Tunnel Diode SoM Transmitters** The hardware-specific fingerprints of transmitters are closely related to its architecture as well as components in the transmit chain. The tunnel diode SoM transmitter employ a novel architecture that incorporates unique hardware blocks, thereby unveiling opportunities for effective fingerprinting. Our experimental observations have revealed that the I-V curve of the tunnel diode exhibits distinct variations from one component to another. These variations, stemming from inherent imperfections in the components, potentially leave unique and identifiable fingerprints in the emitted signals.

**Unknown Device Detection in Radiometric Fingerprinting** In radiometric fingerprinting system implementations, data-driven ML classifiers trained with fingerprints from enrolled devices, are used for identification. One realistic attack model involves adversaries using unenrolled devices. Detecting such devices requires the classifier to identify data points that differ from its training data. This is particularly challenging due to the lack of prior knowledge of unknown fingerprints. The solution to address this issue depends on the choice of classifier. In Paper IV, we briefly discuss solutions for the SVM classifier, and another collaborative work proposes a solution for the neural network classifier [108]. Still, designing a comprehensive system that effectively manages unenrolled devices is an open challenge.
Obfuscating Fingerprints to Defense Privacy Threats Radiometric fingerprinting can be used implicitly by overhearing the wireless communication signals. This implies that fingerprinting is a double-edged sword. On the one hand, it can enhance security for legitimate entities by adding an additional identification layer without the typical overhead associated with such operations. On the other hand, this technology could potentially be exploited by illegitimate entities to track users, raising significant privacy and security concerns [109]. To mitigate this privacy threat, finding a practical and effective method to obfuscate hardware-specific fingerprints remains an unresolved research area.

Leveraging Fingerprinting to Improve Communication Quality Hardware imperfections typically exist due to the limitation of manufacturing processing as well as the compromise between cost and performance. For example, while manufacturers could employ high-quality components with high precision in their radios, this approach would raise the cost per device. The tolerance of acceptable hardware imperfections depends on the radio technology parameters, such as data rates, operating frequencies, and modulation schemes. 5G cellular technology, operating at millimeter-wave frequencies with GHz bandwidths, is particularly sensitive to such imperfections. Hardware imperfections can significantly degrade the quality of the communication. By analyzing transmitter fingerprints, the receiver can adapt its signal processing algorithms specifically to that transmitter. This tailored approach can lead to clear signal reception and reduced error rates.
Trådlös kommunikation är en viktig förutsättning för IoT. Utvecklingen går mot allt mindre IoT-enheter och därmed även ett minskande utrymme för batterier och andra energikällor. De relativt begränsade energiresurserna tvingar fram nya strömsnål och robusta kommunikationslösningar. Denna avhandling adresserar detta behov och presenterar nya radiosändare med extremt låg effekt. Vidare utvecklar vi i avhandlingen passiva och robusta radiometriska fingeravtryckssystem för identifiering av olika sändare med lågt effektbehov.


Radiometriska fingeravtryck som identifierar trådlösa sändare är baserade på fabriktypiska radioenheter och inneboende små men normala ofullkomligheter i hårdvaran. Fingeravtrycket tillför därför ett värdefullt säkerhetslager genom en identifiering av sändaren utan att kräva en kostsam aktiv kommunikation mellan enheter. Metoden skyddar trådlösa system från säkerhetsShot inklusive kloning av enheter, förfalskningsattacker och replay-attacker.
I avhandlingen undersöks tillämpningen av fingeravtryck både på konventionella radiosändare och backscatter-sändare. Fokus är anpassning av fingeravtrycksmetoden för verkliga driftsättningar.


Tidigare fingeravtrycksmetoder kräver sofistikerad signalbearbetning som behöver avsevärd a datorresurser. Vårt arbete demonstrerar att det är möjligt att designa en resurseffektiv fingeravtryckskedja. Allt från signalinhämtning till extraktion och klassificering av radiofaktorerna har integrerats sömlöst inom ramen för en enda SoC. Detta förenklar storskalig användning av fingeravtryckssystem på billiga standardiserade IoT-enheter.


Avhandlingen är positionerad i området strömsnåla IoT-enheter som är ett snabbt växande. Vårt vetenskapliga bidrag är framförallt förståelsen för hårdvarukomponenter och kanalfaktorers roll i kommunikationssystemet. Vi visar också de praktiska konsekvenserna av dessa insikter i verkliga tillämpningar.
Appendix A.
Radio Datasets

The following datasets are used to compare the energy per bit and communication range between commodity radios, backscatter radios, and the Judo system introduced in this dissertation.
<table>
<thead>
<tr>
<th>Chip</th>
<th>Radio Standard</th>
<th>Year</th>
<th>Bitrate (bps)</th>
<th>Tx Power (dBm)</th>
<th>Power Draw (mW)</th>
<th>EPB (nJ/bit)</th>
</tr>
</thead>
<tbody>
<tr>
<td>UBI206 [110]</td>
<td>WiFi</td>
<td>2022</td>
<td>6M</td>
<td>-</td>
<td>20</td>
<td>3.3</td>
</tr>
<tr>
<td>UBI206</td>
<td>WiFi</td>
<td>2022</td>
<td>72.2M</td>
<td>-</td>
<td>20</td>
<td>0.3</td>
</tr>
<tr>
<td>DA16200 [111]</td>
<td>WiFi</td>
<td>2019</td>
<td>1M</td>
<td>9.5</td>
<td>280.5</td>
<td>280.5</td>
</tr>
<tr>
<td>DA16200</td>
<td>WiFi</td>
<td>2019</td>
<td>6M</td>
<td>8</td>
<td>280.5</td>
<td>46.8</td>
</tr>
<tr>
<td>DA16200</td>
<td>WiFi</td>
<td>2019</td>
<td>54M</td>
<td>2</td>
<td>231</td>
<td>4.3</td>
</tr>
<tr>
<td>ESP32 [112]</td>
<td>WiFi</td>
<td>2016</td>
<td>1M</td>
<td>19.5</td>
<td>792</td>
<td>792</td>
</tr>
<tr>
<td>ESP32</td>
<td>WiFi</td>
<td>2016</td>
<td>54M</td>
<td>16</td>
<td>627</td>
<td>11.6</td>
</tr>
<tr>
<td>CC3200 [113]</td>
<td>WiFi</td>
<td>2014</td>
<td>1M</td>
<td>18</td>
<td>1000.8</td>
<td>1000.8</td>
</tr>
<tr>
<td>CC3200</td>
<td>WiFi</td>
<td>2014</td>
<td>1M</td>
<td>13</td>
<td>615.6</td>
<td>615.6</td>
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<tr>
<td>CC3200</td>
<td>WiFi</td>
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<td>6M</td>
<td>18</td>
<td>914.4</td>
<td>152.4</td>
</tr>
<tr>
<td>CC3200</td>
<td>WiFi</td>
<td>2014</td>
<td>6M</td>
<td>4</td>
<td>504</td>
<td>84</td>
</tr>
<tr>
<td>CC3200</td>
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<td>2014</td>
<td>54M</td>
<td>14.8</td>
<td>824.4</td>
<td>15.3</td>
</tr>
<tr>
<td>CC3200</td>
<td>WiFi</td>
<td>2014</td>
<td>54M</td>
<td>4</td>
<td>504</td>
<td>9.3</td>
</tr>
<tr>
<td>CYW43362 [114]</td>
<td>WiFi</td>
<td>2010</td>
<td>11M</td>
<td>18.5</td>
<td>1152</td>
<td>104.7</td>
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<tr>
<td>CYW43362</td>
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<td>2010</td>
<td>54M</td>
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<td>CYW43362</td>
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<td>936</td>
<td>14.4</td>
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<td>CYW43438 [115]</td>
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<td>1M</td>
<td>20</td>
<td>1152</td>
<td>1152</td>
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<td>CYW43438</td>
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<td>54M</td>
<td>15</td>
<td>936</td>
<td>17.3</td>
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<td>EFR32FG28</td>
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<td>35.82</td>
<td>17.9</td>
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<td>Apollo4 [116]</td>
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<td>2020</td>
<td>2M</td>
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<td>17.16</td>
<td>8.6</td>
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<tr>
<td>nRF5340 [118]</td>
<td>BLE</td>
<td>2019</td>
<td>1M</td>
<td>0</td>
<td>12.3</td>
<td>12.3</td>
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<tr>
<td>nRF5340</td>
<td>BLE</td>
<td>2019</td>
<td>2M</td>
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<td>12.6</td>
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<tr>
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<td>9</td>
<td>4.5</td>
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<tr>
<td>CC2652R [121]</td>
<td>BLE</td>
<td>2018</td>
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<td>nRF52840 [123]</td>
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<td>250K</td>
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<td>SX1276 [128]</td>
<td>LoRa</td>
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<td>4.8K</td>
<td>20</td>
<td>396</td>
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<td>4.8K</td>
<td>7</td>
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<td>RN2483 [129]</td>
<td>LoRa</td>
<td>2015</td>
<td>10.9K</td>
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<td>ATA8520E [130]</td>
<td>SigFox</td>
<td>2016</td>
<td>100</td>
<td>14</td>
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</table>

Table 1. Power consumption, data rate and energy per bit of commercial radios. Estimating the power consumption of radios can be challenging when measurements cannot be directly performed on the hardware. Power consumption measurements highly depend on the scenarios, including parameter configurations (data rate, transmission power, and duty cycling) and hardware setups (CPU, memory, peripherals, clock, and power management unit status). The measurement scenarios vary between different chips. Most datasheets report the power consumption when the power management unit is enabled. Some datasheets only list the power consumption with the maximum transmission power. When multiple transmission power levels or data rates are available, several combinations are listed for a comprehensive range of power efficiency calculations.
Table 2. Power consumption, data rate and energy per bit of backscatter radios.

When multiple data rates are available and the power estimation setup is not clearly stated, the maximum data rate is considered for energy per bit calculation. The power consumption reported in different works depends on the implementation. * represents the work covered in this dissertation.

<table>
<thead>
<tr>
<th>Backscatter System</th>
<th>Year</th>
<th>Bitrate (bps)</th>
<th>Implementation</th>
<th>Power Draw (μW)</th>
<th>EPB (nJ/bit)</th>
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</thead>
<tbody>
<tr>
<td>Judo*</td>
<td>2022</td>
<td>100K</td>
<td>PCB prototype-RF analog components</td>
<td>48.00</td>
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<tr>
<td>SyncScatter [10]</td>
<td>2021</td>
<td>300K</td>
<td>IC prototype 65nm</td>
<td>30.00</td>
<td>0.06</td>
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<tr>
<td>TunnelScatter [48]</td>
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<td>2.9K</td>
<td>PCB prototype-RF analog components</td>
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<tr>
<td>PLoRa [13]</td>
<td>2018</td>
<td>6.25K</td>
<td>Simulation FPGA</td>
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<td>35.20</td>
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<td>LoRa Backscatter [12]</td>
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<td>37.5K</td>
<td>Simulation IC 65nm</td>
<td>9.25</td>
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<td>3K</td>
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<td>FS Backscatter [66]</td>
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<td>Simulation HSPICE</td>
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<td>Simulation IC 65nm</td>
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<td>1M</td>
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<td>Simulation - modeling</td>
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</table>

Table 3. Communication range of backscatter radios. The communication range of a backscatter tag is determined by the product of emitter-tag distance $d_1$ and tag-receiver distance $d_2$. * represents the work covered in this dissertation.
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CC2538, 32-bit Arm Cortex-M3 Zigbee, 6LoWPAN, and IEEE 802.15.4 wireless MCU with 512kB Flash and 32kB RAM. https://www.ti.com/product/CC2538.


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